

# 可见光通信网络信道容量和干扰管理

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# Introduction

- Use the visible light as the information carrier, LED as the transmitter and Photon Detector(PD) as the receiver;
- Vast license-free frequency spectrum;
- No electromagnetic interference;
- Inherent security;
- High data rate, 96 Mbps in IEEE 802.15.7, Gbps boost data rates in some research.

# Introduction

## VLC signal features

- Intensity modulation and direct detection  $\Rightarrow$  **real and non-negative** ;
  - Eye safety  $\Rightarrow$  **peak optical power (amplitude) limited** ;
  - Practical illumination  $\Rightarrow$  **average optical power limited**.
- 
- Shannon Formula and Gaussian distribution cannot be used for VLC.

# Motivation

## Opening Problem of VLC: capacity and input distribution

- The capacity-achieving distribution of VLC is **discrete over a finite set of points**<sup>1</sup>
- The number of the discrete points, the amplitude of the discrete points and the probabilities of the discrete points are unknown;
- **Exhaustive Search**, without analytical expression.<sup>1,2</sup>
- Some recent literatures have spent efforts on investigation how to approximate the channel capacity of VLC, i.e., upper bounds or lower bounds;

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<sup>1</sup>J. G. Smith, “The information capacity of amplitude- and variance- constrained scalar gaussian channels,” Inf. Contr., vol. 18, no. 3, pp. 203-219, Feb. 1971.

<sup>2</sup>T. Chan, S. Hranilovic, and F. Kschischang, “Capacity-achieving probability measure for conditionally gaussian channels with bounded inputs,” IEEE Trans. Inf. Theory, vol. 51, no. 6, pp. 2073–2088, Jun 2005.

# Motivation

## Upper bound of VLC Capacity

- Duality-based approach<sup>3</sup>
- Sphere packing method<sup>4</sup>
- Sphere packing method<sup>5</sup>
- Sphere packing based Recursive approach<sup>6</sup>

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<sup>3</sup>A. Lapidoth, S. M. Moser, and M. Wigger, “On the capacity of free-space optical intensity channels,” *IEEE Trans. Inf. Theory*, vol. 55, no. 10, pp. 4449–4461, Oct. 2009.

<sup>4</sup>A. A. Farid and S. Hranilovic, “Capacity bounds for wireless optical intensity channels with gaussian noise,” *IEEE Trans. Inf. Theory*, vol. 56, no. 12, pp. 6066–6077, Dec. 2010.

<sup>5</sup>J. B. Wang, Q. S. Hu, J. Wang, M. Chen, and J. Y. Wang, “Tight bounds on channel capacity for dimmable visible light communications,” *J. Lightw. Technol.*, vol. 31, no. 23, pp. 3771–3779, Dec. 2013.

<sup>6</sup>A. Chaaban, J. Morvan, and M. Alouini, “Free-space optical communications: capacity bounds, approximations, and a new sphere-packing perspective,” *IEEE Trans. Commun.*, vol. 64, no. 3, pp. 1176–1191, Mar. 2016.

# Motivation

## Lower bound of VLC Capacity

- Differential entropy maximization (Peak and average optical power)<sup>7</sup>
- Discrete entropy maximization (Average optical power)<sup>8</sup>
- Discrete entropy maximization (Peak and average optical power)<sup>9</sup>

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<sup>7</sup>A. Lapidoth, S. M. Moser, and M. Wigger, “On the capacity of free-space optical intensity channels,” IEEE Trans. Inf. Theory, vol. 55, no. 10, pp. 4449–4461, Oct. 2009.

<sup>8</sup>A. A. Farid and S. Hranilovic, “Capacity bounds for wireless optical intensity channels with gaussian noise,” IEEE Trans. Inf. Theory, vol. 56, no. 12, pp. 6066–6077, Dec. 2010.

<sup>9</sup>A. A. Farid and S. Hranilovic, “Channel capacity and non-uniform signaling for free-space optical intensity channels,” IEEE J. Sel. Areas Commun., vol. 17, no. 9, pp. 1553–1563, Dec. 2009.

# Exact Capacity of VLC Channel

- System Model
- Inexact Gradient Projection Method
- Simulation Results
- Conclusion



# Exact Capacity of VLC Channel

## IM/DD Channel Model

$$Y = X + Z, \quad (2.1)$$

$$0 \leq X \leq A, \quad (2.2)$$

$$\mathbb{E}\{X\} \leq \mu, \quad (2.3)$$

$$Z \sim \mathcal{C}(0, \sigma^2). \quad (2.4)$$

## $P(X)$ – the distribution of $X$

$$\Pr\{X = x_k\} = p_k \geq 0, \quad \forall k \in \mathcal{K}, \quad (2.5a)$$

$$\mathbb{E}\{X\} = \sum_{k=1}^K p_k x_k \leq \mu, \quad (2.5b)$$

$$\sum_{k=1}^K p_k = 1, \quad (2.5c)$$

$$0 \leq x_k \leq A, \quad \forall k \in \mathcal{K}, \quad (2.5d)$$

# Exact Capacity of VLC Channel

## Capacity

$$C^{\text{VLC}} = \max_{P(X)} I(X; Y) \quad (2.6a)$$

$$= \max_{P(X)} h(Y) - h(Y|X) \quad (2.6b)$$

$$= \max_{P(X)} - \int_{-\infty}^{\infty} f_Y(y) \log_2 f_Y(y) dy - \frac{1}{2} \log_2 2\pi e \sigma^2, \quad (2.6c)$$

$f_Y(y)$  –the probability density function (PDF) of  $Y$

$$f_Y(y) = \frac{1}{\sqrt{2\pi}\sigma} \sum_{k=1}^K p_k e^{-\frac{(y-x_k)^2}{2\sigma^2}} \quad (2.7)$$

# Problem Formulation

Finding the exact capacity of VLC channel can be formulated as the following optimization problem:

$$\min_{K, \{p_k\}, \{x_k\}} \int_{-\infty}^{\infty} \frac{\sum_{k=1}^K p_k e^{-\frac{(y-x_k)^2}{2\sigma^2}}}{\sqrt{2\pi}\sigma} \log_2 \frac{\sum_{j=1}^K p_j e^{-\frac{(y-x_j)^2}{2\sigma^2}}}{\sqrt{2\pi}\sigma} dy \quad (2.8)$$

$$\text{s.t. } \Pr \{X = x_k\} = p_k \geq 0, \forall k \in \mathcal{K}, \quad (2.9)$$

$$\mathbb{E} \{X\} = \sum_{k=1}^K p_k x_k \leq \mu, \quad (2.10)$$

$$\sum_{k=1}^K p_k = 1, \quad (2.11)$$

$$0 \leq x_k \leq A, \forall k \in \mathcal{K}, \quad (2.12)$$

# Concise form

## Variables

$$\mathbf{x} \triangleq [x_1, \dots, x_k, \dots, x_K]^T,$$

$$\mathbf{p} \triangleq [p_1, \dots, p_k, \dots, p_K]^T,$$

$$\phi(\mathbf{p}) \triangleq \int_{-\infty}^{\infty} \mathbf{p}^T \mathbf{q}(y) \log_2 \mathbf{p}^T \mathbf{q}(y) dy,$$

$$\mathbf{q}(y) \triangleq \frac{1}{\sqrt{2\pi}\sigma} \left[ e^{-\frac{(y-x_1)^2}{2\sigma^2}}, \dots, e^{-\frac{(y-x_k)^2}{2\sigma^2}}, \dots, e^{-\frac{(y-x_K)^2}{2\sigma^2}} \right]^T.$$

## Exact capacity of VLC

$$\min_{K, \mathbf{p}, \mathbf{x}} \phi(\mathbf{p}) \quad (2.13a)$$

$$\text{s.t. } \mathbf{p}^T \mathbf{1} = 1, \quad (2.13b)$$

$$\mathbf{p}^T \mathbf{x} \leq \mu, \quad (2.13c)$$

$$\mathbf{p} \geq \mathbf{0}, \quad (2.13d)$$

# Optimization of $K$ and $\mathbf{x}$

## Fixing $\mathbf{x}$ with Equal Spacing

Pick the  $K$  values  $\{x_l\}_{1 \leq l \leq K}$  from the range  $[0, A]$  with equal spacing, i.e.,

$$x_l \triangleq \frac{A}{K-1} (l-1), \quad l = 1, \dots, K. \quad (2.14)$$

**Proposition 1:** Suppose that  $K^*$  and  $\{x_k^*\}_{1 \leq k \leq K^*}$  are the optimal solutions of problem (2.13), where  $x_1^* \leq \dots \leq x_k^* \leq x_{k+1}^* \leq \dots \leq x_{K^*}^*$ . Then, for a given  $\varepsilon_0 > 0$ , when  $K \geq \left\lceil \frac{A}{\varepsilon_0} \right\rceil + 1$ , there exists an equal spacing sequence  $\{x_l\}_{1 \leq l \leq K}$  satisfying

$$\min_{1 \leq l \leq K} |x_k^* - x_l| \leq \varepsilon_0, \quad \forall k \in \mathcal{K}^*, \quad (2.15)$$

where  $\mathcal{K}^* \triangleq \{1, 2, \dots, K^*\}$ .

# Optimization of $\mathbf{p}$

## Convex Problem with Simplex Constraints

$$\min_{\mathbf{p}} \phi(\mathbf{p}) \quad (2.16a)$$

$$\text{s.t. } \mathbf{p} \in \Upsilon, \quad (2.16b)$$

where  $\Upsilon \triangleq \{\mathbf{p} | \mathbf{p}^T \mathbf{1}_K = 1, \mathbf{p}^T \mathbf{x} \leq \mu, \mathbf{p} \geq \mathbf{0}, \}$ .

Gradient Projection Method = Gradient Descent + feasible Projection

# Optimization of $\mathbf{p}$

## Gradient Projection Method

For  $n$ th Iteration,

$$\hat{\mathbf{p}}_{n+1} \triangleq \mathbf{p}_n - \alpha_n \nabla \tilde{\phi}(\mathbf{p}_n), \quad (2.17a)$$

$$\mathbf{p}_{n+1} = \text{Proj}_{\Upsilon}(\hat{\mathbf{p}}_{n+1}), \quad (2.17b)$$

where  $\alpha_n \in (0, 1]$  is a stepsize of the  $n$ th iteration,

$$\nabla \phi(\mathbf{p}) \triangleq \int_{-\infty}^{\infty} \left( \mathbf{q}(y) \log_2 \mathbf{p}^T \mathbf{q}(y) + \frac{1}{\ln 2} \mathbf{q}(y) \right) dy, \quad (2.18)$$

$$\text{Proj}_{\Upsilon}(\mathbf{x}) = \begin{cases} \mathbf{x}, & \text{if } \mathbf{x} \in \Upsilon; \\ \arg \min_{\hat{\mathbf{x}} \in \Upsilon} \|\mathbf{x} - \hat{\mathbf{x}}\|^2, & \text{otherwise.} \end{cases} \quad (2.19)$$

Neither the objective function  $\phi(\mathbf{p})$  nor the gradient  $\nabla \phi(\mathbf{p})$  has an analytic expression.

# Optimization of $\mathbf{p}$

## Approximation by the Numerical Integration

$$\tilde{\phi}(\mathbf{p}) \triangleq \int_{-\tau_1}^{A+\tau_1} \mathbf{p}^T \mathbf{q}(y) \log_2 \mathbf{p}^T \mathbf{q}(y) dy, \quad (2.20a)$$

$$\nabla \tilde{\phi}(\mathbf{p}) \triangleq \int_{-\tau_2}^{A+\tau_2} \left( \mathbf{q}(y) \log_2 \mathbf{p}^T \mathbf{q}(y) + \frac{1}{\ln 2} \mathbf{q}(y) \right) dy. \quad (2.20b)$$

where  $\tau_1 > 0$  and  $\tau_2 > 0$ .

Approximation error can be arbitrarily small with a sufficiently large  $\tau_1$  and  $\tau_2$ .



# Inexact Gradient Descent Method

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## Algorithm 1 Inexact Gradient Descent Method.

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**Initialization:** choose  $K \geq 2$ ,  $\lambda_{K-1} \leq 0$ , set  $c_2, c_3$  as the stopping parameter.  
 set  $n = 0$ , and choose a random starting point  $\mathbf{p}_0 \in \Upsilon$ ;

1: **repeat**

2:   **repeat**

3:     Update  $\tilde{\phi}(\mathbf{p}_n)$  by (2.17a) and  $\nabla \tilde{\phi}(\mathbf{p}_n)$  by (2.17b);

4:     Compute stepsize  $\alpha_n$  by Algorithm 1;

5:     Update  $\mathbf{p}_{n+1} = \text{Proj}_{\Upsilon}(\mathbf{p}_n - \alpha_n \nabla \tilde{\phi}(\mathbf{p}_n))$ ;

6:     Let  $n \leftarrow n + 1$ ;

7:   **until**  $\|\tilde{\mathbf{p}}_{n+1} - \mathbf{p}_n\| \leq c_2$ ;

8:    $\lambda_K = \tilde{\phi}(\mathbf{p}_n)$ ;

9: **until**  $|\lambda_K - \lambda_{K-1}| \leq c_3$ ;

**Output:**  $\mathbf{p}^{\text{opt}} = \mathbf{p}_n$ ,  $K^{\text{opt}} = K$ .

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# Backtracking Line Search

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## Algorithm 2 Backtracking Line Search

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**Initialization:** choose  $\bar{\alpha}, \rho, c \in (0, 1)$ ;

1: **while**  $\tilde{\phi}(\bar{\mathbf{p}}_{n+1}) \leq \tilde{\phi}(\mathbf{p}_n) + c\bar{\alpha}\nabla\tilde{\phi}(\mathbf{p}_n)^T(\bar{\mathbf{p}}_{n+1} - \mathbf{p}_n)$ , where  $\bar{\mathbf{p}}_{n+1} =$   
 $\text{Proj}_{\Upsilon}[\mathbf{p}_n - \bar{\alpha}\nabla\tilde{\phi}(\mathbf{p}_n)]$  **do**

2:  $\bar{\alpha} \leftarrow \rho\bar{\alpha}$ ;

3: **end while**

**Output:**  $\alpha_n = \bar{\alpha}$ .

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# Optimality

## Theorem 1: (Objective Function)

The inexact objective function  $\tilde{\phi}(\mathbf{p})$  can approach the objective function  $\phi(\mathbf{p})$  with an arbitrary small error, i.e.,

$$\left| \phi(\mathbf{p}) - \tilde{\phi}(\mathbf{p}) \right| \leq \varepsilon_1. \quad (2.21)$$

where  $\varepsilon_1 > 0$  is a given accuracy. Mathematically, for a given  $\varepsilon_1 > 0$ , there exists a large parameter  $\tau_1 > \sigma$  satisfying

$$\max \left\{ \left| 2B_1 \operatorname{erfc} \left( \frac{\tau_1}{\sigma\sqrt{2}} \right) + \frac{\tau_1 e^{-\frac{\tau_1^2}{2\sigma^2}}}{\sqrt{2\pi}\sigma \ln 2} \right|, \left| 2B_2 \operatorname{erfc} \left( \frac{\tau_1}{\sigma\sqrt{2}} \right) + \frac{(\tau_1 + 2A) e^{-\frac{\tau_1^2}{2\sigma^2}}}{\sqrt{2\pi}\sigma \ln 2} \right| \right\} \leq \varepsilon_1, \quad (2.22)$$

where  $B_1 \triangleq \frac{1+2\ln(\sigma\sqrt{2\pi})}{4\ln 2}$ ,  $B_2 \triangleq \frac{\sigma^2+A^2+2\sigma^2\ln(\sigma\sqrt{2\pi})}{\sigma^2 4\ln 2}$ .

# Optimality

## Theorem 2: (Gradient Function)

The inexact gradient function  $\nabla\tilde{\phi}(\mathbf{p})$  can approach the gradient function  $\nabla\phi(\mathbf{p})$  with an arbitrarily small gap, i.e., we have

$$\left\| \nabla\phi(\mathbf{p}) - \nabla\tilde{\phi}(\mathbf{p}) \right\| \leq \varepsilon_2. \quad (2.23)$$

where a given accuracy  $\varepsilon_2 > 0$ . Mathematically, for a given  $\varepsilon_2 > 0$ , there exists a  $\tau_2 \geq \sigma$  satisfying

$$\sqrt{K} \left( \max \left\{ \left| 2B_1 \operatorname{erfc} \left( \frac{\tau_2}{\sigma\sqrt{2}} \right) + \frac{\tau_2 e^{-\frac{\tau_2^2}{2\sigma^2}}}{\sqrt{2\pi}\sigma \ln 2} \right|, \right. \right. \\ \left. \left| 2B_2 \operatorname{erfc} \left( \frac{\tau_2}{\sigma\sqrt{2}} \right) + \frac{(\tau_1 + 2A) e^{-\frac{\tau_2^2}{2\sigma^2}}}{\sqrt{2\pi}\sigma \ln 2} \right| \right\} + \frac{\operatorname{erfc} \left( \frac{\tau_2}{\sigma\sqrt{2}} \right)}{\ln 2} \right) \leq \varepsilon_2. \quad (2.24)$$

## Theorem 3: (Convergence Analysis)

For a given  $K$ ,  $\{\mathbf{p}_n\}$  converges to the optimal solution of problem (2.16), and  $\{\phi(\mathbf{p}_n)\}$  converges to the corresponding optimal value of problem (2.16).

# Simulation Settings

## Different Methods

- exhaustive search<sup>10</sup>
- entropy maximization based non-uniform distribution<sup>11</sup>
- Blahut-Arimoto algorithm<sup>12,13</sup>

## Simulation Parameter

- $\kappa \triangleq \frac{A}{\mu}$ ,  $\text{SNR} \triangleq \frac{\mu}{\sigma}$
- $\sigma^2 = 1$ ,  $Z \sim \mathcal{N}(0, 1)$

<sup>10</sup>J. G. Smith, “The information capacity of amplitude- and variance- constrained scalar gaussian channels,” Inf. Contr., vol. 18, no. 3, pp. 203-219, Feb. 1971.

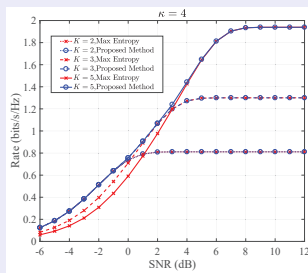
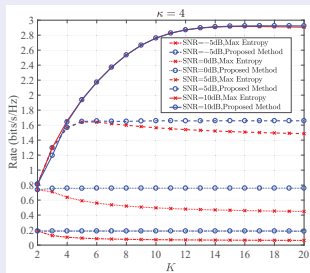
<sup>11</sup>A. A. Farid and S. Hranilovic, “Channel capacity and non-uniform signaling for free-space optical intensity channels,” IEEE J. Sel. Areas Commun., vol. 17, no. 9, pp. 1553–1563, Dec. 2009.

<sup>12</sup>R. Blahut, “Computation of channel capacity and rate-distortion functions,” IEEE Trans. Inf. Theory, vol. 18, no. 4, pp. 460–473, Jul. 1972.

<sup>13</sup>J. H. G. Dauwels., On Graphical Models for Communications and Machine Learning: Algorithms, Bounds, and Analog Implementation, ETH PhD thesis, 2006.

# Simulation Results

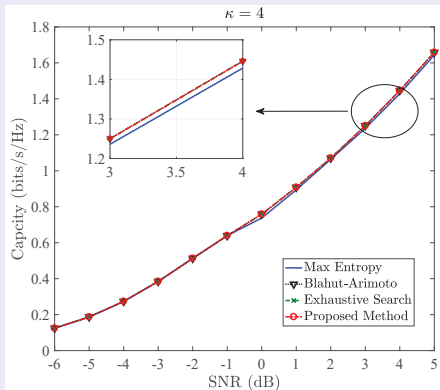
## Capacity versus $K$ or SNR



- Inexact Gradient Projection: Higher rate for same  $K$  and SNR
- Max-Entropy: rate versus  $K$  first increases and then decreases
- As SNR increases, the gap of rate decreases.
- As SNR increases, a large  $K$  is needed to achieve the capacity.

# Simulation Results

## Capacity versus SNR



# Simulation Results

## Computing Timing

SNR(dB)	CPU time (sec.)		
	Exhaustive Search	Blahut-Arimoto	Proposed Method
-2	2.05	3	1.66
0	809.72	1393.91	51.11
2	16801.8	5820.94	444.08
4		41540.72	542.02

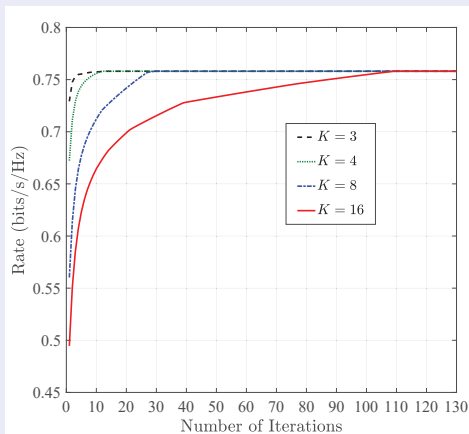
$\kappa = 4$ , MATLAB<sup>®</sup> (R2016b) with 3.4GHz CPUs and 16GB RAM

- Significantly less than the other
- More slowly increases



# Convergence

Rate versus the Number of Iterations ( $\kappa = 4$ , SNR = 0dB)



- Converge to the rate of the optimal number of points.
- The redundant points can be removed by optimizing  $\mathbf{p}$ .

# Optimal input distribution

## Optimal input distribution( $\kappa = 4$ )

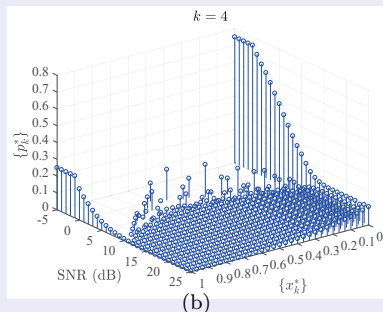
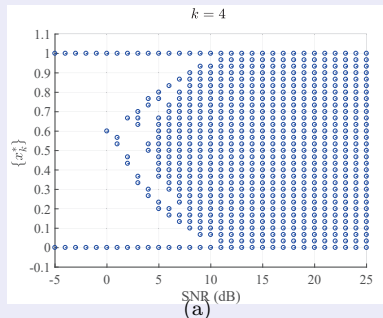


Figure 2.1: (a) The optimal input positions  $\{x_k^*\}$  versus SNR for  $\kappa = 4$ ; (b) The optimal input distribution  $\{x_k^*, p_k^*\}$  versus SNR for  $\kappa = 4$ .

The revealed optimal discrete input distribution is the OOK modulation at low SNR, and the PAM at high SNR.

# Conclusion

For achieving channel capacity of VLC,

- based on the inexact gradient projection , we develop a low-complexity method which can converge to the optimal input distribution under both peak and average optical power constraints;
- we prove theoretically the approximations of objection function and its gradient can be arbitrarily accurate by numerical integration;
- we prove theoretically the optimality the inexact gradient projection method and provide numerical results to verify the results;
- we numerically verify hat our method is the most efficient by comparing it with existing methods.

S. Ma, R. Yang, Y. He, S. Lu, F. Zhou, N. Al-Dhahir, and S. Li, “Achieving channel capacity of visible light communication,” IEEE Syst. J., pp. 1-12, Mar. 2020.(early access)

# Achievable Rate with Closed-form for VLC

- VLC System Model
- ABG Lower Bound
- Simulation Results

# VLC System Model

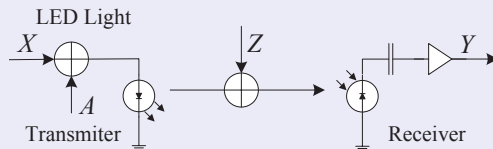


Figure 3.1: The schematic of VLC SISO System.

- IM/DD SISO system
- $X \sim p_X$
- The parameter  $A > 0$  is the fixed DC bias to guarantee the transmitted signal  $X + A \geq 0$
- $Y = X + A + Z$ , where  $Z \sim \mathcal{N}(0, \sigma^2)$

# VLC System Model

- Amplitude constraint:

$$-A \leq X \leq A \quad (3.1)$$

- Average optical power:

$$\mathbb{E}\{X\} = 0, P_o = \mathbb{E}\{X + A\} = A, \quad (3.2)$$

- Average electrical power:

$$\mathbb{E}\{X^2\} = \varepsilon, P_e = \mathbb{E}\{(X + A)^2\} = \varepsilon + A^2. \quad (3.3)$$

# Continuous input distribution

The input  $X$  follows continuous distribution and  $p_X = f_X(x)$  is the probability density function of  $X$ , such that

$$\begin{cases} \int_{-A}^A f_X(x) dx = 1, \\ f_X(x) \geq 0, x \in [-A, A], \\ f_X(x) = 0, x \notin [-A, A], \end{cases} \quad (3.4)$$

Moreover, the distribution  $f_X(x)$  satisfies

$$\mathbb{E}\{X\} = \int_{-A}^A x f_X(x) dx = 0, \quad (3.5a)$$

$$\mathbb{E}\{X^2\} = \int_{-A}^A x^2 f_X(x) dx = \varepsilon. \quad (3.5b)$$

# Achievable Rate

$$C^{\text{SISO}} \geq \max_{f_X(x)} I(X; Y) \quad (3.6a)$$

$$= \max_{f_X(x)} h(Y) - h(Z) \quad (3.6b)$$

$$= \max_{f_X(x)} h(X + A + Z) - h(Z) \quad (3.6c)$$

$$\geq \max_{f_X(x)} \frac{1}{2} \log_2 \left( 2^{2h(X)} + 2^{2h(Z)} \right) - h(Z) \quad (3.6d)$$

$$= \max_{f_X(x)} \frac{1}{2} \log_2 \left( 1 + \frac{2^{2h(X)}}{2\pi e \sigma^2} \right), \quad (3.6e)$$



# Achievable Rate

- The optimization problem of maximum differential entropy

$$\min_{f_X(x)} \int_{-A}^A f_X(x) \ln f_X(x) dx \quad (3.7a)$$

$$\text{s.t.} \int_{-A}^A f_X(x) dx = 1, \quad (3.7b)$$

$$\int_{-A}^A x f_X(x) dx = 0, \quad (3.7c)$$

$$\int_{-A}^A x^2 f_X(x) dx = \varepsilon. \quad (3.7d)$$

- The differential entropy maximization problem (3.7) is convex.
- We can solve problem (3.7) by lagrangian multiplier method

# Achievable Rate

## theorem 3.1 (ABG lower bound)

For VLC SISO channel, the ABG lower bound of the channel capacity is given by

$$C^{\text{SISO}} \geq \frac{1}{2} \log_2 \left( 1 + \frac{e^{1+2(\alpha+\gamma\varepsilon)}}{2\pi\sigma^2} \right), \quad (3.8)$$

lower bound can be achieved by  $X$  whose distribution is given by

$$f_X(x) = \begin{cases} e^{-1-\alpha-\beta x-\gamma x^2}, & -A \leq x \leq A; \\ 0, & \text{otherwise.} \end{cases} \quad (3.9)$$

# ABG Lower Bound

The parameters  $\alpha$ ,  $\beta$  and  $\gamma$  be the solutions of the following equations

$$T(A) - T(-A) = e^{1+\alpha}, \quad (3.10a)$$

$$\beta \left( e^{A(\beta-\gamma A)} - e^{-A(\beta+\gamma A)} - e^{1+\alpha} \right) = 0, \quad (3.10b)$$

$$e^{A(\beta-\gamma A)} \left( (\beta - 2\gamma A) e^{-2A\beta} - \beta - 2\gamma A \right) + (\beta^2 + 2\gamma) e^{1+\alpha} = 4\gamma^2 \varepsilon e^{1+\alpha}, \quad (3.10c)$$

where  $T(x) \triangleq \sqrt{\pi} \frac{e^{\frac{\beta^2}{4\gamma}} \operatorname{erf}\left(\frac{\beta+2\gamma x}{2\sqrt{\gamma}}\right)}{2\sqrt{\gamma}}.$

- For convenience, we define parameter  $\phi \triangleq \frac{A^2}{\varepsilon}.$
- Note that, when  $\phi = 3$ , the proposed ABG input distribution is the uniform distribution.

# Simulation Results

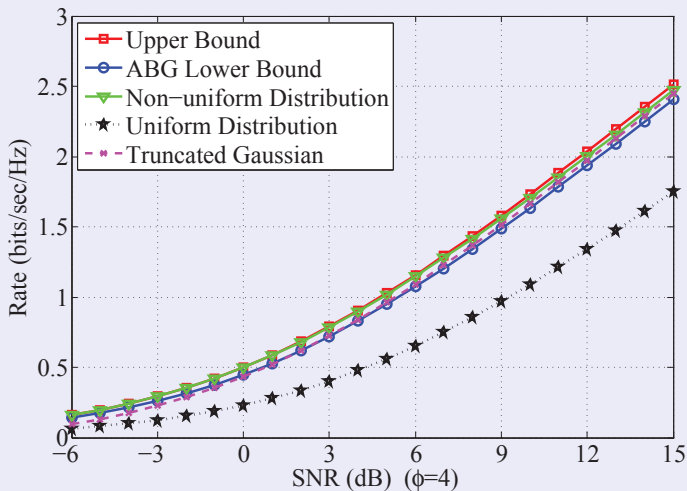


Figure 3.2: The achievable rate (bits/sec/Hz) versus SNR (dB) with  $\phi = 4$ .

# Optimal Power Allocation for Mobile NOMA VLC

- System Model of NOMA VLC
- Achievable Rates for Mobile Users of NOMA
- Optimal Power Allocation Scheme for Mobile Users of NOMA

# System Model

## Downlink of NOMA VLC

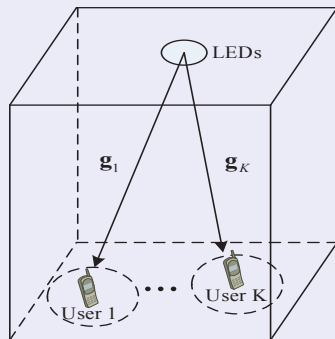


Figure 4.1: The schematic of downlink of NOMA VLC.

- Assume that the locations area of users can be bounded by circles;

## The geometry of mobile users

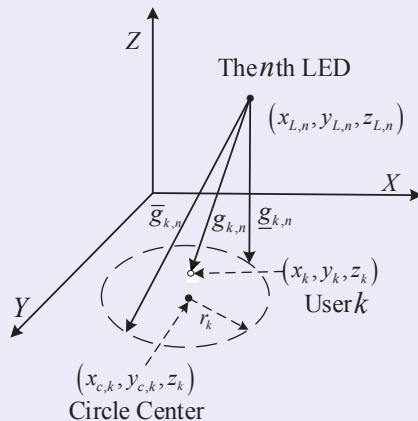


Figure 4.2: The geometry of mobile users

## Channel Model of Mobile Users

- The user  $k$  is located within the circular area with center  $(x_{c,k}, y_{c,k}, z_k)$  and radius  $r_k$ , i.e.,

$$\mathcal{R}_k \triangleq \left\{ (x_k, y_k) \mid 0 \leq (x_k - x_{c,k})^2 + (y_k - y_{c,k})^2 \leq r_k^2 \right\}. \quad (4.1)$$

- Then, the channel gain between the  $n$ th LED and user  $k$  is given as

$$g_{k,n} = \frac{(m+1) A_R}{2\pi d_{k,n}^2} \left( \frac{z_{L,n} - z_k}{d_{k,n}} \right)^{m+1} = \frac{\beta (z_{L,n} - z_k)^{m+1}}{d_{k,n}^{m+3}}, \quad (4.2)$$



## Channel Model of Mobile Users

- Furthermore, the CSI uncertainty vector  $\Delta \mathbf{g}_k$  can be characterized by an ellipsoidal region:

$$\mathcal{G}_k \triangleq \{ \Delta \mathbf{g}_k | \Delta \mathbf{g}_k^T \mathbf{C}_k \Delta \mathbf{g}_k \leq v_k \}, \quad (4.3)$$

where  $\mathbf{C}_k = \mathbf{C}_k^T \succeq \mathbf{0}$  controls the extension of the ellipsoid

- Before user  $k$  decoding the signal  $s_i$ , the residual received signal is given as

$$y_{k,i}^{\text{Mob,SIC}} = \mathbf{g}_k^T \mathbf{w}_i s_i + \sum_{l=1}^{i-1} \Delta \mathbf{g}_k^T \mathbf{w}_l s_l + \sum_{m=i+1}^K \mathbf{g}_k^T \mathbf{w}_m s_m + \mathbf{g}_k^T \mathbf{b} + n_k, \quad (4.4)$$

## Achievable Rate of Mobile Users

- The lower bound of  $R_{k,i}^{\text{Mob}}$  is given by

$$R_{k,i}^{\text{Mob}} = \max_{\{f_i(x_i)\}} I(x_i; y_{k,i}^{\text{Mob,SIC}}) \quad (4.5a)$$

$$= \max_{\{f_i(x_i)\}} h(y_{k,i}^{\text{Mob,SIC}}) - h(y_{k,i}^{\text{Mob,SIC}} | x_i) \quad (4.5b)$$

$$\geq \frac{1}{2} \log_2 \frac{2\pi\sigma_k^2 + \sum_{j=i}^K |\mathbf{g}_k^T \mathbf{w}_j|^2 \tau_j + \sum_{l=1}^{i-1} |\Delta \mathbf{g}_k^T \mathbf{w}_l|^2 \tau_l}{2\pi \left( \sigma_k^2 + \sum_{m=i+1}^K |\mathbf{g}_k^T \mathbf{w}_m|^2 \varepsilon_m + \sum_{n=1}^i |\Delta \mathbf{g}_k^T \mathbf{w}_n|^2 \varepsilon_n \right)}, \quad (4.5c)$$

where  $\tau_j \triangleq e^{1+2(\alpha_j + \gamma_j \varepsilon_j)}$

# Optimal Power Allocation for Mobile Users

## Problem Formulation

- The optimal power allocation problem of mobile users can be formulated as

$$\min_{\{\mathbf{w}_k\}_{k=1}^K} \sum_{k=1}^K \varepsilon_k \|\mathbf{w}_k\|^2 \quad (4.35a)$$

$$\text{s.t. } R_{k,i}^{\text{L,Mob}} \geq \bar{R}_i, 1 \leq i \leq k \leq K, \quad (4.35b)$$

$$\sum_{k=1}^K A_k |w_{k,n}| \leq b, \forall n \in \mathcal{N}, \quad (4.35c)$$

$$\mathbf{g}_k = \hat{\mathbf{g}}_k + \Delta \mathbf{g}_k, \Delta \mathbf{g}_k \in G_k, \forall k \in \mathcal{K}. \quad (4.35d)$$

# Optimal Power Allocation for Mobile Users

## Optimal Power Allocation

- Let define  $\mathbf{W}_k = \mathbf{w}_k \mathbf{w}_k^T$ , and then constraint (4.35b) can be reformulated as

$$\Delta \mathbf{g}_k^T \widehat{\mathbf{W}}_{k,i} \Delta \mathbf{g}_k + 2 \Delta \mathbf{g}_k^T \widetilde{\mathbf{W}}_{k,i} \widehat{\mathbf{g}}_k + \widehat{\mathbf{g}}_k^T \widetilde{\mathbf{W}}_{k,i} \widehat{\mathbf{g}}_k \geq c_{k,i}, \quad (4.36)$$

where

$$\widehat{\mathbf{W}}_{k,i} \triangleq \sum_{j=1}^K \tau_j \mathbf{W}_j - 2^{2\bar{R}_i+1} \pi \sum_{n=1}^i \varepsilon_n \mathbf{W}_n - 2^{2\bar{R}_i+1} \pi \Gamma_{k,i} \sum_{m=i+1}^K \varepsilon_m \mathbf{W}_m, \quad (4.37)$$

$$\widetilde{\mathbf{W}}_{k,i} \triangleq \sum_{j=i}^K \tau_j \mathbf{W}_j - 2^{2\bar{R}_i+1} \pi \Gamma_{k,i} \sum_{m=i+1}^K \varepsilon_m \mathbf{W}_m, \quad (4.38)$$

$$c_{k,i} \triangleq \left( 2^{2\bar{R}_i} - 1 \right) 2\pi \sigma_k^2. \quad (4.39)$$

# Optimal Power Allocation for Mobile Users

## SDR

- Furthermore, by ignoring the rank-one constraint for all  $\mathbf{W}_k$ 's, the SDR of problem (4.40) is given by

$$\min_{\{\mathbf{W}_k\}_{k=1}^K} \varepsilon_k \text{Tr}(\mathbf{W}_k) \quad (4.40a)$$

$$\begin{aligned} \text{s.t. } & \Delta \mathbf{g}_k^T \widehat{\mathbf{W}}_{k,i} \Delta \mathbf{g}_k + 2 \Delta \mathbf{g}_k^T \widetilde{\mathbf{W}}_{k,i} \widehat{\mathbf{g}}_k + \widehat{\mathbf{g}}_k^T \widetilde{\mathbf{W}}_{k,i} \widehat{\mathbf{g}}_k \geq c_{k,i}, \\ & 1 \leq i \leq k \leq K, \end{aligned} \quad (4.40b)$$

$$\sum_{k=1}^K \text{Tr}(\mathbf{W}_k \mathbf{e}_n \mathbf{e}_n^T) \leq \frac{b^2}{A_k^2}, \forall n \in \mathcal{N}, \quad (4.40c)$$

$$\mathbf{W}_k \succeq \mathbf{0}, \forall k \in \mathcal{K}, \quad (4.40d)$$

$$\Delta \mathbf{g}_k \in \mathcal{G}_k, \forall k \in \mathcal{K}. \quad (4.40e)$$

# Optimal Power Allocation for Mobile Users

## S-Lemma

- By using  $\mathcal{S}$ -lemma, we obtain the following conservative approximation of problem (4.40) as follows

$$\max_{\{\mathbf{W}_k\}} \sum_{k=1}^K \varepsilon_k \text{Tr}(\mathbf{W}_k) \quad (4.41a)$$

$$\text{s.t.} \quad \begin{bmatrix} \widehat{\mathbf{W}}_{k,i} + \lambda_{k,i} \mathbf{C}_k & \widetilde{\mathbf{W}}_{k,i} \widehat{\mathbf{g}}_k, \\ \widehat{\mathbf{g}}_k^T \widetilde{\mathbf{W}}_{k,i} & \widehat{\mathbf{g}}_k^T \widetilde{\mathbf{W}}_{k,i} \widehat{\mathbf{g}}_k - c_{k,i} - \lambda_{k,i} v_k \end{bmatrix} \succeq \mathbf{0},$$

$$1 \leq i \leq k \leq K, \quad (4.41b)$$

$$\sum_{k=1}^K \text{Tr}(\mathbf{W}_k \mathbf{e}_n \mathbf{e}_n^T) \leq \frac{b^2}{A_k^2}, \forall n \in \mathcal{N}, \quad (4.41c)$$

$$\mathbf{W}_k \succeq \mathbf{0}, \forall k \in \mathcal{K}, \quad (4.41d)$$

which is a convex SDP.

# Simulation Results

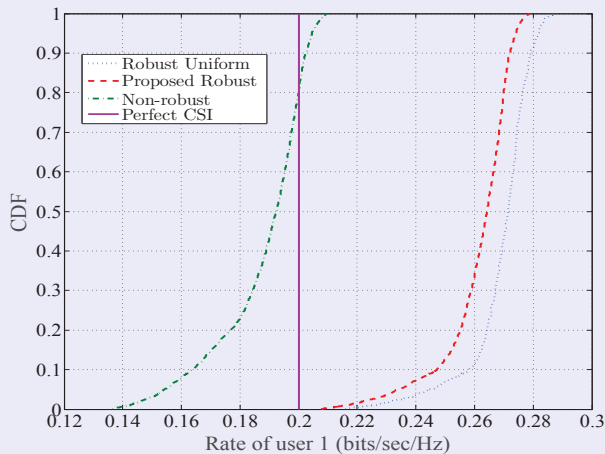


Figure 4.3: CDF of rate over 20000 random channel realizations of user 1.

# Simulation Results

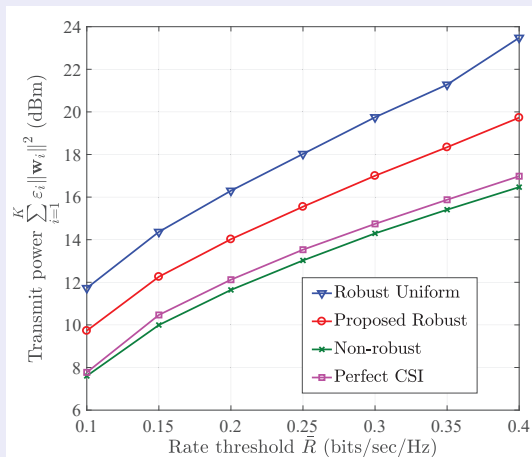


Figure 4.4: Transmit power  $\sum_{i=1}^K \varepsilon_i \|\mathbf{w}_i\|^2$  (dBm) versus rate threshold  $\bar{R}$ .



# Simulation Results

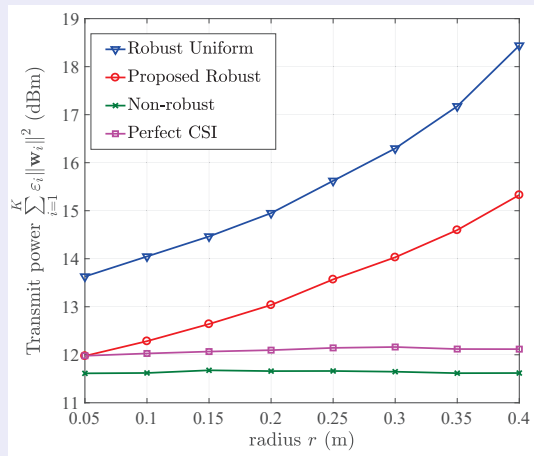


Figure 4.5: Transmit power  $\sum_{i=1}^K \varepsilon_i \|\mathbf{w}_i\|^2$  (dBm) versus radius  $r$  (m).

# Summary

## Summary

- We proposed an efficient inexact gradient descent method to obtain the channel capacity-achieving input distribution of VLC.
- We developed a closed-form ABG lower bound for VLC channels;
- We proposed a robust power allocation scheme for mobile users scenario in NOMA VLC networks.

# Published papers



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Thanks For Your Attention!